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13/7/92

Introduction



Feed positioning system controls the rotation of the feed turret to enable operation at any of the selected frequency band. Earlier reports outlined various schemes and performance specifications. (Feed Positioning system NCRA Technical reports, Accn# 90198. & Feed Drive control System ITR/SERVO/1 20/8/91.)

This report, describes hardware and software for the working prototype.

Encoder

Encoder selected for the application is ROD 420 - 2048 manufactured by Heidenhain. This is an incremental encoder with 2048 line counts per revolution of the encoder shaft. It gives two incremental output signals with a 90° phase shift. Direction of rotation determines whether channel A or channel B leads. In addition, a reference pulse is produced once per revolution. All signals are balanced and conform to RS - 422 specifications. Further information and data sheet is included in the Appendix. 1.

Mechanical Configuration

Feed drive motor is coupled to the feed turret through two stages of reduction (Fig. 1) A worm gear with 300 :1 reduction and feed bull gear with a reduction of 5:1. Encoder is coupled to the output shaft of the worm gear. ie there is a reduction of 5:1 between the encoder shaft and the feed turret.

Controller Hardware

Please refer to Fig.2 and Fig.3 for the description that follows;

The controller is based on a microcontroller (Intel 80C51FA). Regulated 5 V DC is supplied to the encoder from the controller board. Outputs of the encoder are connected to balanced line receivers. Use of balanced drivers and receivers enable long cable

lengths with excellent common mode noise immunity. TTL outputs of the line receivers are further passed through digital filters to reject noise spikes. Incremental signals are then decoded into COUNT and DIRECTION signals. COUNT gives a pulse train at twice the line count. DIRECTION indicates the direction of rotation. These two signals are connected to the microcontroller input pins. Position count is maintained by incrementing or decrementing a 16 bit counter on the 80C51FA. This is done in hardware by enabling the appropriate mode. Thus, there is no danger of missed counts.

Reference pulse output goes to an interrupt on the 80C51. This reference pulse occurs at fixed shaft position of the encoder and is very accurate. The 16 bit position count is initialised on the first occurrence of the reference pulse. The position count maintained is thus with respect to the first index pulse. Since there is a 5:1 reduction between the encoder shaft and the feed turret, there will be multiple reference pulses in one complete rotation of the feed turret. It is thus necessary to 'mark' one of the reference pulses. This 'marked' reference can then be used to maintain absolute position count. This marking can be provided by a limit switch. Positioning accuracy of the limit switch is not important as the count is maintained from the reference pulse.

Feeds will be positioned at 90° intervals on the turret. This corresponds to 450° on the encoder shaft. 450° rotation of the encoder shaft corresponds to 2048×1.25 lines. This corresponds to 5120 ($2048 \times 1.25 \times 2$) pulses at the microcontroller. Feeds 1 to 4 are assigned position counts differing by 5120.

Pulse width modulation (PWM) is used to control the power to the motor. In this scheme, power to the motor is switched ON and OFF at a fast rate. The ratio of ON to OFF times determines the average power fed to the motor. Speed of the motor can thus be controlled by controlling the duty cycle of the switches. MOSFETS (IRF 540) are used as the switching elements. These are capable of handling 10 A safely and have a turn on resistance of only $85\text{ m}\Omega$ when driven properly.

PWM controller is also implemented in hardware in the 80C51FA. It is only necessary to write the appropriate count to the proper

register after suitable configuration. H Bridge is used to control the Servo motor. Four MOSFET switches and servo motor are connected in the form of an 'H'. Direction of rotation can be controlled by turning ON switch pairs 1 & 4 or 2 & 3. Thus, speed of the servo motor is controlled by PWM and direction by choosing the appropriate pair of MOSFETS.

Brake of the servo motor is controlled by an output pin of 80C51 through an opto isolator and a Darlington transistor. Interlocking is done to ensure that the motor does not turn ON if the brake is not released. Short circuit / stalled motor current limit can be implemented by choosing appropriate resistance at the ground end of the H bridge. When voltage drop across the resistance exceeds, 0.6 V transistor Q is turned ON. This turns OFF the MOSFETS as the charge pump cannot supply enough gate drive. Charge pump generates an isolated 12 - 16 DC voltage source capable of driving the MOSFET gates. Gates of the MOSFETS can thus be driven by a source which is about 12 V higher than the source and drain. This is required to ensure proper turn ON of the MOSFETS. The charge pump oscillator frequency is at about 50 - 100 KHz.

Controller Software

Present software on the controller is very basic and ensures standalone operation with a PC or any RS 232 / RS 422 terminal. Presently, the servo hand held terminal (Oriole PBT) is used.

Counts corresponding to the four feeds are put in the EPROM. These are 90° apart on the feed turret. Position count is with respect to the first reference pulse encountered. Feed one is assumed to be located at the first reference pulse location.

After the first reference pulse, position counter is initialised and the motor is stopped. When any of the feeds is 'selected' the direction and magnitude of the displacement required is computed. This is used to choose the appropriate direction and PWM for the motor. PWM is increased gradually to limit the start up current. As long as the target position is outside the proportional region, motor runs at full speed. When target position is near, speed is gradually reduced. Brake is applied when position error is below a certain limit. Currently, proportional

region starts when the target position is within 256 counts. Brake is applied when error is less than 5 counts. These parameters have to be modified based on performance with feeds mounted.

There is also a 'manual' mode where the motor can be 'Run', 'Stopped' and the 'Direction' changed. Position count by obtained by pressing the 'Position' key. Appendix.2 lists the current commands supported.

Appendix.3 gives the assembly program listing. Program is straight forward and easy to follow with the comments. Initialisation and proper use of hardware resources is unambiguous.

Further improvements & modifications

HARDWARE

1. Currently, provision for connecting the 'marking' limit switch is not there. This limit switch can be connected to int1 pin of 80C51 through a pull up resistor. Noise rejection is not critical as, position accuracy of the switch is not important. Limit switch is sensed only at the start. Int1 should be disabled after initialisation.
2. If the FPS is to be used as an MCM, an RS 485 interface is required. ie. only one twisted pair is used. In such a case, transmitter (U8A) should be enabled only when FPS has to transmit. This can be achieved by connecting the enable pin of U8A to P1.6 or P1.7 of 80C51. One of the LED's will then indicate the transmit status.
3. Power ground (PGND) and signal ground (GND) should be separate for best noise immunity and isolation. Oriole hand held terminal may misbehave when large switching spikes occur on the common ground.
4. As indicated on the schematic, filter capacitors must be placed close to the servo motor.
5. It is suggested to restrict the maximum current to the motor to 10 A eventhough the rated maximum is 12 A. This can be achieved

by restricting the 50 V power supply to 10 A. Max.

6. Heat sinks for Q1 - Q4 should be improved for reliable continuous operation. Commercial heatsinks like AFCOSET should be used. Insulation is not required between any two transistors of the top or bottom.

SOFTWARE :

7. When starting from rest PWM should be increased GRADUALLY. Full power should never be applied when the motor is at rest.

8. For use as an MCM, commands have to be embedded in a proper frame to make them compatible with other MCM's.

9. Automatic address recognition feature of 80C51 can be used to simplify the software.

10. Appropriate TX enable (P1.6 or P1.7) should be added depending on the hardware change to interface with other MCM's.

11. Proportional range may have to be increased to enable smooth braking at the final position without over shoot. Count at which brake has to be applied can be determined only after study with the feeds loaded on the turret. PWM can be decremented rapidly to ensure faster braking.

12. Direction of rotation can be reversed after zero PWM has been reached to realise sudden braking.

13. Speed and displacement can be monitored to detect motor stall and other fault conditions.

14. Positions of the feeds can be downloaded from the ABC for flexibility. Similarly target position can also be downloaded.

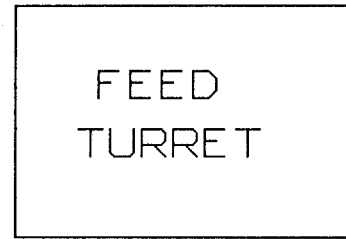
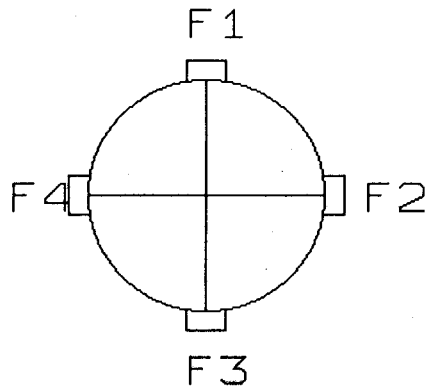
15. Watchdog timer on 80C51 can be activated for reliability.

16. Feeds can be given multiple position counts to enable selection of shortest path between any two feeds subject to cable wrap.

Some of the suggestions are illustrated in Appendix 4

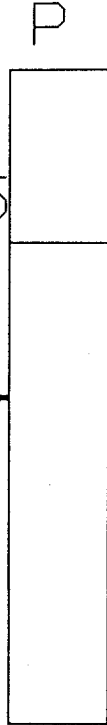
References

1. Vaidya.V, Kulkarni.G.R., "Feed Drive C
ITR/SERVO/1 20/8/91
2. A.Ramakrishna., "Feed Positioning Syst
NCRA Technical Reports Accn # 90198.
3. Intel, 8 Bit Microcontroller Handbook.
4. Motorola, Power MOSFET Handbook.

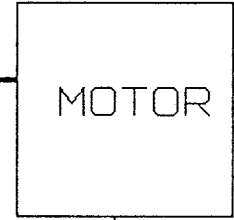


1 : 5

FEED
BULL
GEAR



1 : 300



ENCODER

ENCODER

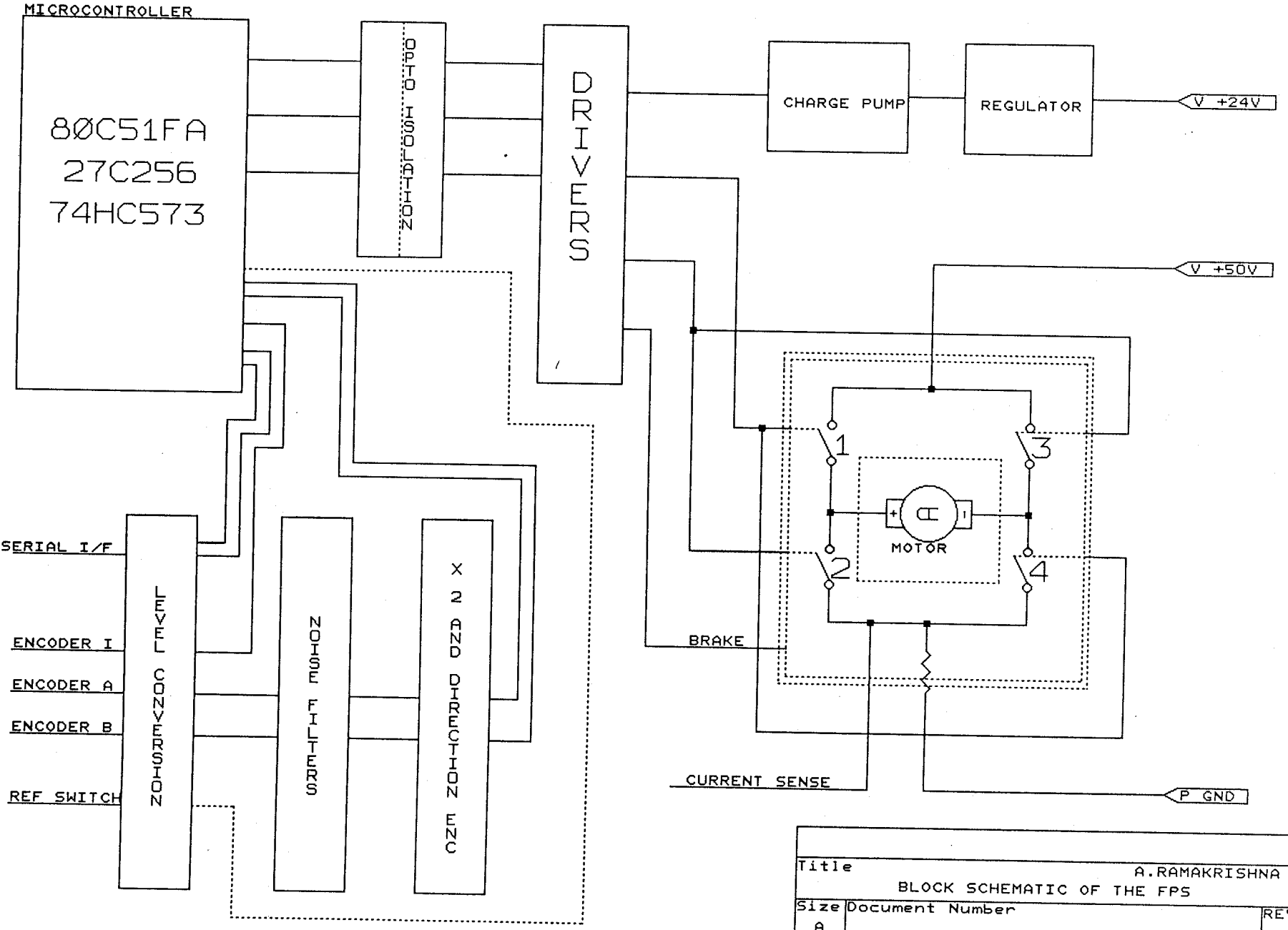
MOTOR
CONTROLLER

P = PINION

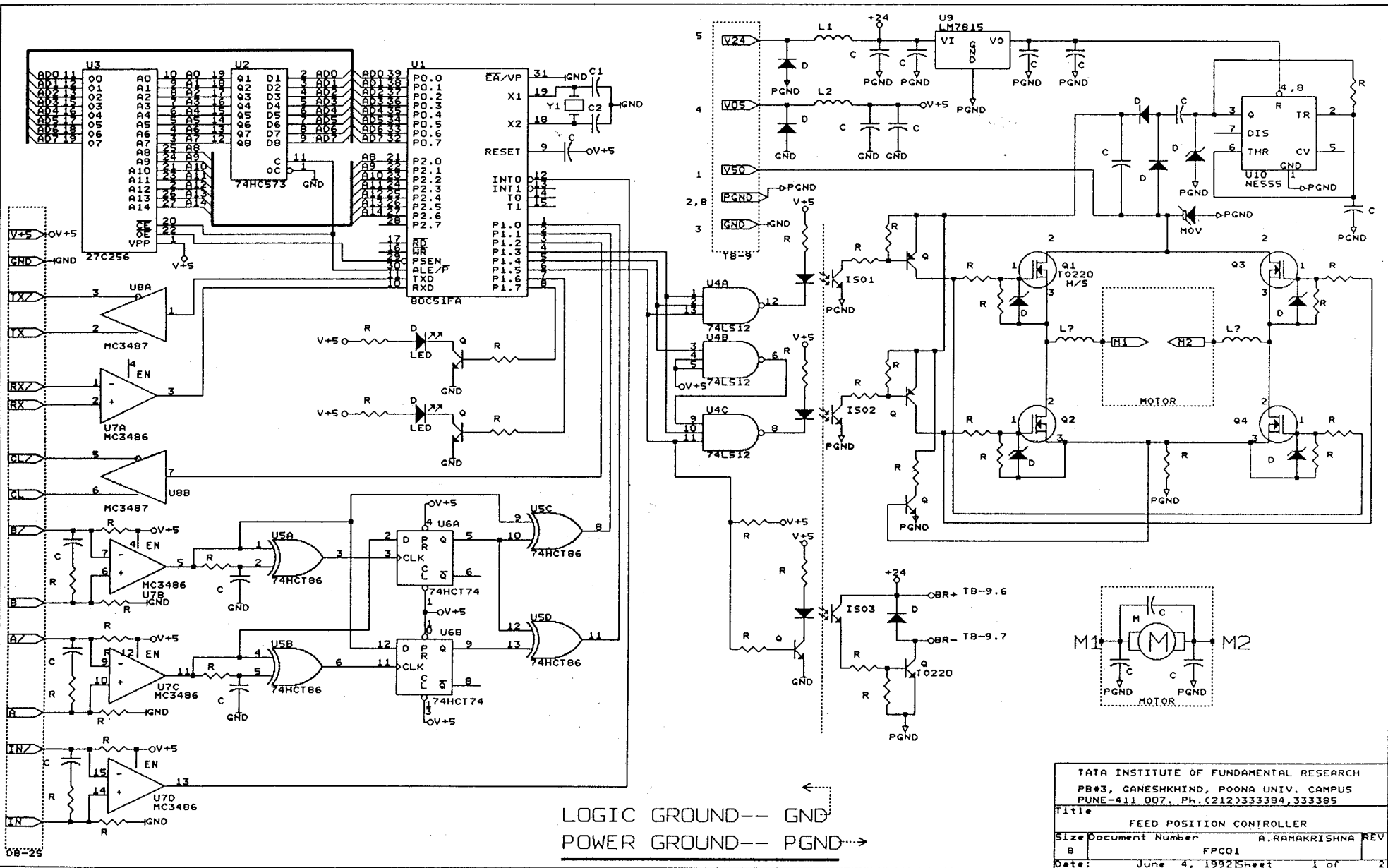
B = BRAKE

Adopted from Ref. 1

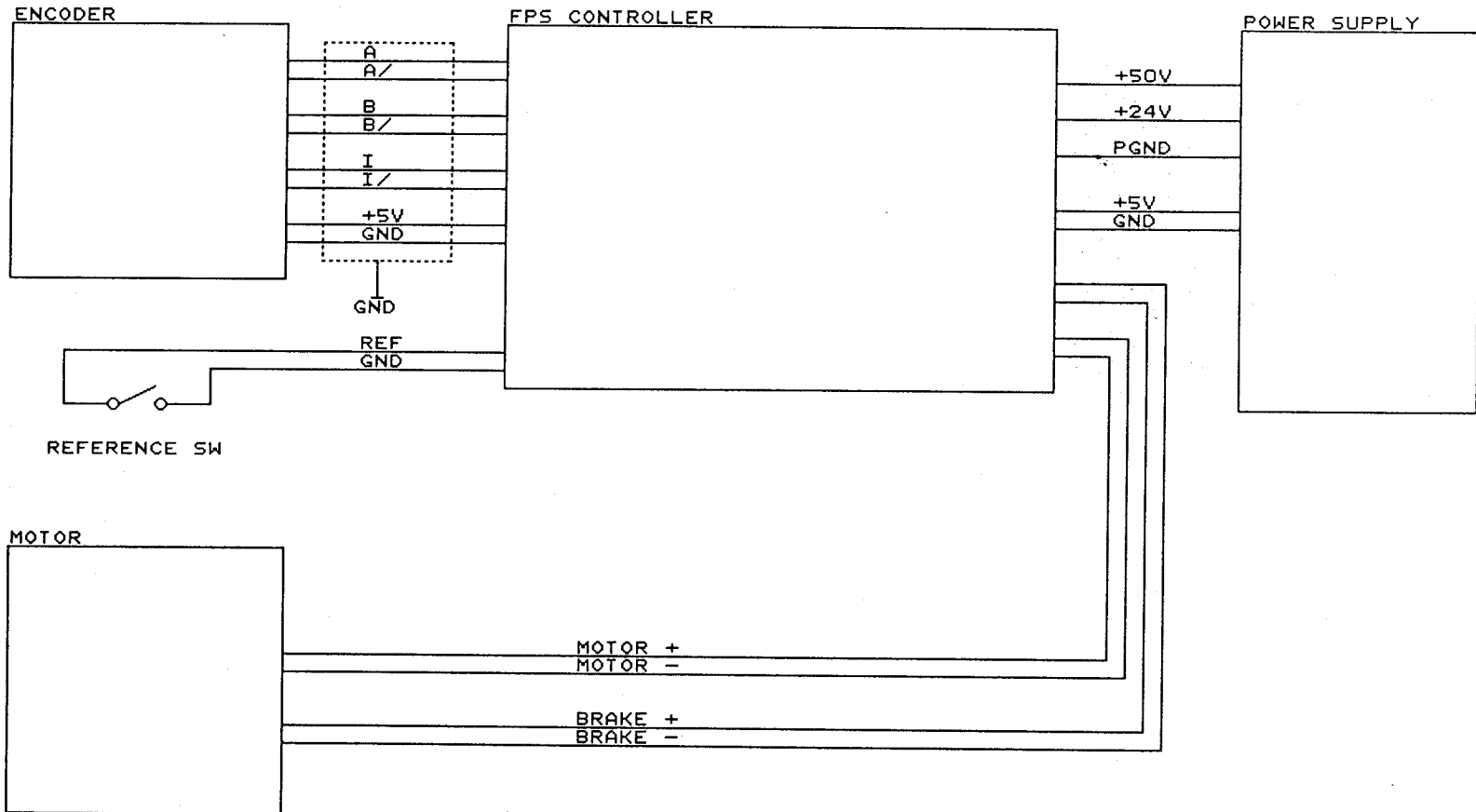
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MECHANICAL CONFIGURATION			
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BLOCK SCHEMATIC OF THE FPS			
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Date:	July 11, 1992	Sheet	of



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Title	FEED POSITION CONTROLLER
Size	Document Number A.RAMAKRISHNA REV
B	FPC01
Date:	June 4, 1992 Sheet 1 of 2



Title		
INTERCONNECTION FPS		
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Technical Specifications

Mechanical Data

ROD 420

Line counts	50/60/100/120/125/128/150/180/200/250/254/256/360/400/420/500/512/600/625/635/720/800/ 900/1000/1024/1080/1125/1250/1270/1500/1750/1800/2000/2048/2080/2500/2540/2920/3000/ 3600/3750/4000/4096/4500/5000 (special line counts upon request)	
Accuracy	$\pm 18'' / z$ ($\triangleq 1/20$ grating period)	$z = \text{line count}$
Resolution	0.018'' with 5000 lines and 4-fold evaluation in the subsequent electronics	
Slewing speed	max. 6 000 rpm	
Moment of inertia of rotor	$1.8 \cdot 10^{-6} \text{ kgm}^2$	
Torque at 20° C (68° F)	$\leq 0.01 \text{ Nm}$	
Shaft load	axial	max. 40 N
	radial	max. 60 N (at shaft end)
Weight	approx. 0.3 kg (0.55 lb)	
Type of protection	IP 64 according to IEC 529	
Operating temperature	0° to 70° C (32° to 158° F)	
Storage temperature	-30° to 80° C (-22° to 176° F)	
Vibration (10 to 2000 Hz)	$\leq 100 \text{ m/s}^2$	
Shock (11 ms)	$\leq 1000 \text{ m/s}^2$	

Appendix

Technical Specifications



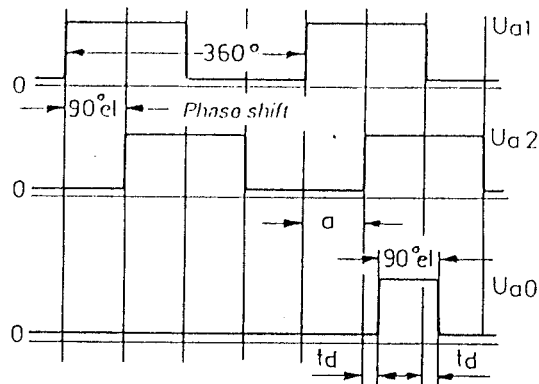
Electrical Data

Power supply

5 V \pm 5 % / max. 205 mA (without load)

Light source: filament lamp 5 V/0.6 W

Output signals



ROD 420

Incremental signals

TTL square-wave pulse trains U_{a1} , U_{a2} and their inverted signals \bar{U}_{a1} and \bar{U}_{a2} . U_{a2} lag U_{a1} with clockwise rotation (see Fig. 1)

Edge separation

$a \geq 0.8 \mu s$ at scanning frequency 160 kHz

Reference signal

1 square-wave pulse U_{a0} per revolution and its inverted signal \bar{U}_{a0}

Lag time

lag of pulse U_{a0} to signals U_{a1} and U_{a2}
 $t_{dl} \leq 50 ns$

Signal level

$U_{aHigh} \geq 2.5 V$ at $-I_{aHigh} = 20 mA$
 $U_{aLow} \leq 0.5 V$ at $I_{aLow} = 20 mA$

Loading

$-I_{aHigh} \leq 20 mA$
 $I_{aLow} \leq 20 mA$
 $C_{Load} \leq 1000 pF$

Switching times

rise time $t_+ \leq 100 ns$
fall time $t_- \leq 100 ns$

Scanning frequency

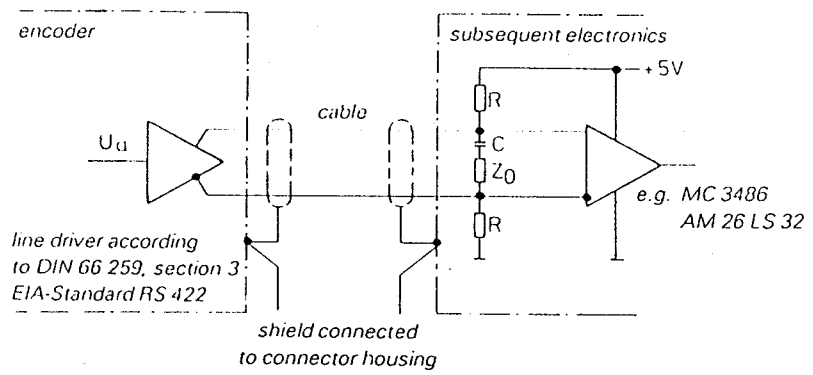
0 to 160 kHz

Slewing speed

0 to $(160/z \times 10^3 \times 60)$ rpm

$z = \text{line count}$

Recommended input circuitry of subsequent electronics



$R = 4.7\text{ k}\Omega$; prevents switching of the receiver during line break

$C = 1\text{ to }10\text{ nF}$; reduces the DC loading of the rotary encoder

$Z_0 = 120\text{ up to }140\ \Omega$ (with HEIDENHAIN cable) or corresponding to the characteristic impedance of the cable

Cable length at encoder to subsequent electronics: 1 m (3.3 ft) other lengths upon request; 50 m (164 ft) max., with HEIDENHAIN cable ($4 \times 2 \times 0.14 + 4 \times 0.5$) mm² with sufficient power supply at the encoder

Pin Layout	B1	V+S	A IN	A $\bar{I}N$	A	\bar{A}	B			Gnd	Gnd	V+S
Pin	1	2	3	4	5	6	7	8	9	10	11	12
Signal	U_{a2}	sensor + 5 V	U_{a0}	U_{a0}	U_{a1}	U_{a1}	free	U_{a2}	shield *	0 V	sensor 0 V	+ 5 V
Color	pink	0.25mm ² blue	red	black	brown	green	/	gray	/	0.25mm ² white/green	0.25mm ² white	0.25mm ² brown/green

* shield is on the connector housing and is connected to pin 9 in the connector

Permissible bending radii of cable

Cable diameter	Permissible bending radius for	
	Repeated bending	Stationary config.
$\varnothing\ 4,5\text{ mm (}.18\text{ in.)}$	$R \geq 50\text{ mm (}2\text{ in.)}$	$R \geq 10\text{ mm (}.4\text{ in.)}$
$\varnothing\ 6\text{ mm (}.24\text{ in.)}$	$R \geq 75\text{ mm (}3\text{ in.)}$	$R \geq 20\text{ mm (}.8\text{ in.)}$
$\varnothing\ 8\text{ mm (}.31\text{ in.)}$	$R \geq 100\text{ mm (}4\text{ in.)}$	$R \geq 40\text{ mm (}1.6\text{ in.)}$

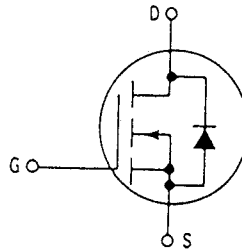
Power Field Effect Transistor

N-Channel Enhancement-Mode

Silicon Gate TMOS

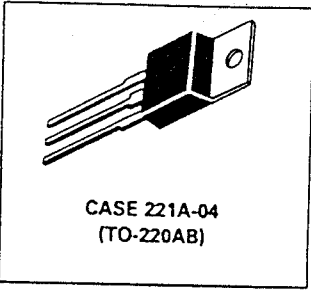
These TMOS Power FETs are designed for low voltage, high speed power switching applications such as switching regulators, converters, solenoid and relay drivers.

Silicon Gate for Fast Switching Speeds
Low $r_{DS(on)}$ to Minimize On-Losses. Specified at Elevated Temperature
Rugged — SOA is Power Dissipation Limited
Source-to-Drain Diode Characterized for Use With Inductive Loads



IRF540
IRF541
IRF542

TMOS POWER FETs
24 and 27 AMPERES
 $r_{DS(on)} = 0.085 \text{ OHM}$
60 and 100 VOLTS
 $r_{DS(on)} = 0.11 \text{ OHMS}$
100 VOLTS



3

MAXIMUM RATINGS

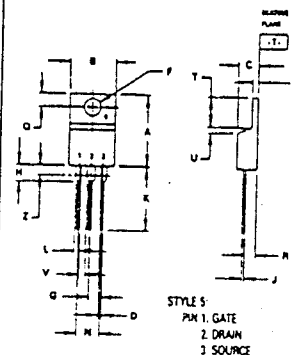
Rating	Symbol	IRF			Unit
		540	541	542	
Drain-Source Voltage	V_{DSS}	100	60	100	Vdc
Drain-Gate Voltage ($R_{GS} = 20 \text{ k}\Omega$)	V_{DGR}	100	60	100	Vdc
Gate-Source Voltage	V_{GS}	± 20			Vdc
Drain Current Continuous, $T_C = 25^\circ\text{C}$ $T_C = 100^\circ\text{C}$ Peak, $T_C = 25^\circ\text{C}$	I_D	27 17 108	24 15 96		Adc
Total Power Dissipation (@ $T_C = 25^\circ\text{C}$ Derate above 25°C)	P_D	125 1			Watts W/°C
Operating and Storage Temperature Range	T_J, T_{stg}	-55 to 150			°C

THERMAL CHARACTERISTICS

Thermal Resistance — Junction to Case — Junction to Ambient	$R_{\theta JC}$ $R_{\theta JA}$	1 62.5	°C/W
Maximum Lead Temp. for Soldering Purposes, 1/8" from Case for 5 Seconds	T_L	300	°C

See the MTP25N10 Designer's Data Sheet for a complete set of design curves for the product on this data sheet.

OUTLINE DIMENSIONS



- NOTES
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982
 2. CONTROLLING DIMENSION: INCH
 3. DIM Z DEFINES A ZONE WHERE ALL BODY AND LEAD IRREGULARITIES ARE ALLOWED.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	14.48	15.75	0.570	0.620
B	9.65	10.28	0.380	0.405
C	4.07	4.82	0.160	0.190
D	0.64	0.88	0.025	0.035
F	3.61	3.73	0.142	0.147
G	2.42	2.66	0.095	0.105
H	2.30	3.33	0.110	0.135
J	2.36	2.55	0.014	0.022
K	2.75	4.77	0.100	0.187
L	1.15	1.39	0.045	0.055
M	4.83	5.33	0.190	0.210
Q	2.54	2.54	0.100	0.100
R	2.04	2.75	0.080	0.110
S	1.15	1.39	0.045	0.055
T	5.97	6.47	0.235	0.255
U	2.00	1.77	0.000	0.050
V	1.15	—	0.045	—
Z	—	2.54	—	0.090

ELECTRICAL CHARACTERISTICS ($T_C = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Max	Unit	
OFF CHARACTERISTICS					
Drain-Source Breakdown Voltage ($V_{GS} = 0, I_D = 0.25 \text{ mA}$)	IRF540, IRF542 IRF541	$V_{(BR)DSS}$	100 60	— —	Vdc
Zero Gate Voltage Drain Current ($V_{DS} = \text{Rated } V_{DSS}, V_{GS} = 0$) ($V_{DS} = 0.8 \text{ Rated } V_{DSS}, V_{GS} = 0, T_J = 125^\circ\text{C}$)		I_{DSS}	— —	0.2 1	mAdc
Gate-Body Leakage Current, Forward ($V_{GSF} = 20 \text{ Vdc}, V_{DS} = 0$)		I_{GSSF}	—	100	nAdc
Gate-Body Leakage Current, Reverse ($V_{GSR} = 20 \text{ Vdc}, V_{DS} = 0$)		I_{GSSR}	—	100	nAdc

ON CHARACTERISTICS*

Gate Threshold Voltage ($V_{DS} = V_{GS}, I_D = 0.25 \text{ mA}$)		$V_{GS(th)}$	2	4	Vdc
Static Drain-Source On-Resistance ($V_{GS} = 10 \text{ Vdc}, I_D = 15 \text{ Adc}$)	IRF540, IRF541 IRF542	$r_{DS(on)}$	— —	0.085 0.11	Ohm
On-State Drain Current ($V_{GS} = 10 \text{ V}$) ($V_{DS} \geq 2.3 \text{ Vdc}$) ($V_{DS} \geq 2.6 \text{ Vdc}$)	IRF540, IRF541 IRF542	$I_{D(on)}$	27 24	— —	Adc
Forward Transconductance ($V_{DS} \geq 2.3 \text{ V}, I_D = 15 \text{ A}$) ($V_{DS} \geq 2.6 \text{ V}, I_D = 15 \text{ A}$)	IRF540, IRF541 IRF542	g_{FS}	6.0 6.0	— —	mhos

DYNAMIC CHARACTERISTICS

Input Capacitance	$(V_{DS} = 25 \text{ V}, V_{GS} = 0, f = 1 \text{ MHz})$	C_{iss}	—	1600	pF
Output Capacitance		C_{oss}	—	800	
Reverse Transfer Capacitance		C_{rss}	—	300	

SWITCHING CHARACTERISTICS*

Turn-On Delay Time	$(V_{DD} = 30 \text{ V}, I_D = 15 \text{ Apk}, R_{gen} = 4.7 \text{ Ohms})$	$t_{d(on)}$	—	30	ns
Rise Time		t_r	—	60	
Turn-Off Delay Time		$t_{d(off)}$	—	80	
Fall Time		t_f	—	30	
Total Gate Charge	$(V_{DS} = 0.8 \text{ Rated } V_{DSS}, V_{GS} = 10 \text{ Vdc}, I_D = \text{Rated } I_D)$	Q_g	40 (Typ)	60	nC
Gate-Source Charge		Q_{gs}	17 (Typ)	—	
Gate-Drain Charge		Q_{gd}	23 (Typ)	—	

SOURCE DRAIN DIODE CHARACTERISTICS*

Forward On-Voltage	$(I_S = \text{Rated } I_D, V_{GS} = 0)$	V_{SD}	1.5 (Typ)	2.3(1)	Vdc
Forward Turn-On Time		t_{on}	Limited by stray inductance		
Reverse Recovery Time		t_{rr}	450 (Typ)	—	ns

INTERNAL PACKAGE INDUCTANCE

Internal Drain Inductance (Measured from the contact screw on tab to center of die) (Measured from the drain lead 0.25" from package to center of die)	L_d	3.5 (Typ) 4.5 (Typ)	— —	nH
Internal Source Inductance (Measured from the source lead 0.25" from package to source bond pad)	L_s	7.5 (Typ)	—	

*Pulse Test: Pulse Width $\leq 300 \mu\text{s}$, Duty Cycle $\leq 2.0\%$.

(1) Add 0.1 V for IRF540 and IRF541.

- RUN : ie Release Brake & accelerate gradually
- STOP : ie: Gradually reduce speed & apply brake
- Direction : ie Change direction of rotation
at rest - changes direction for next R
while running reverses direction by
slowing the motor & restores to speed

- Displays Current Position Count in HEX

Position	Feed	Location
Subst Position	Feed 1	1000 H
Position	Feed 2	2400 H
Position	Feed 3	3800 H
"	"	4C00 H

Terminal : Async 9600, n, 1,


```

#####
;#
;# ASSEMBLY LANGUAGE LISTING OF PROGRAM. - A.RAMAKRISHNA - #
;# #
;#####

```

```

$ NOMOD51
NAME FPS

```

```

restart CODE 00H
start0 CODE 30h
TIMER2 CODE 2BH

```

```

$ INCLUDE (RG51FA.PDF)

```

```

CSEG AT EXTIO

```

```

LJMP INXISR

```

```

CSEG AT

```

```

TIMER2

```

```

LJMP T2ISR

```

```

CSEG AT

```

```

restart

```

```

ljmp start

```

```

CSEG AT

```

```

start0

```

```

start:

```

```

clr p1.5 ;brake on
clr ren
mov IP,#00H ;NO PRIORITY
mov IE,#01H ;EN EXT0,EA=0

MOV T2MOD,#01H ;DN COUNT EN
MOV T2CON,#0AH ;T2 UP/DN MODE+DIS T2

MOV RCAP2H,#0H ;LOW LIMIT FOR UP/DN COUNT
MOV RCAP2L,#0H

clr psw.1 ;speed flag
mov pcon,#0h ;3.68MHz
; mov pcon,#80h ;Double BR for 1.8M
mov tmod,#21h ;t1 auto reload,t0
mov th1,#0ffh ;9600 baud
setb TR1 ;Start timer1
setb cr ;PCA on
; setb cf ;
mov cmod,#00h ;*** :-12, 1.2KHz.
mov ccapm0,#42h ;P1.3(cex0) PWM
mov ccap0h,#0ffh ;pwm=0

mov scon,#40h ;mode 1,clr ti,ri.
setb ren ;EN RX
clr ri
SETB EA ;ENABLE INTERRUPTS(EXT0)
SETB P1.4 ;DIR +

rdchs:
jnb ri,$ ;chs aval ?
mov a,sbuf
clr ri
cjne a,#'r',st ;if ne r may be s OR ...

run:
;chs=r. RUN
clr p1.6 ;led2 off

```

```

cjne    a,#0ffh,rdchs    ;if already running ignore

dec --  a
mov     ccap0h,a
call    delay
cjne    a,#0d,dloop      ;increase speed slowly till 100pc
jmp     rdchs            ;running,look for next chs

cjne    a,#'s',dir       ;if not s may be d OR ....

                                ;chs=s. STOP
clr     p1.7              ;led1 off
setb    p1.6              ;led2 on
mov     a,ccap0h
cjne    a,#0d,rdchs      ;not running? skip
inc     a
mov     ccap0h,a
call    delay
cjne    a,#0ffh,iloop    ;till pwm=0.
clr     p1.5              ;Brake
jmp     rdchs            ;look at next chs

cjne    a,#'d',FE1 ;if not d CHECK FOR...

                                ;chs=d.CHANGE DIRECTION.
clr     p1.7              ;led1 off STOP
setb    p1.6              ;led2 on
mov     a,ccap0h         ;GET CURR PWM
cjne    a,#0ffh,ilop    ;pwm ne 0 ? dec pwm

cpl     p1.4              ;change dir
jbc     psw.1,run        ;old speed NE 0
jmp     rdchs            ;look at next chs

p:setb  psw.1             ;flag for restoring speed
inc     a
mov     ccap0h,a
call    delay
cjne    a,#0ffh,ilop    ;till pwm=0
jmp     flip             ;ok to flip now

CJNE    A,#31H,FE2       ;IF NOT 1 CHECK 2,3,4,p
MOV     R0,#10H          ;MSB REF          1000H
MOV     R1,#00H          ;LSB REF
CALL    TARGET

CJNE    A,#32H,FE3       ;IF NOT 2 CHECK 3,4,p
MOV     R0,#24H          ;MSB REF          2400H
MOV     R1,#00H          ;LSB REF
CALL    TARGET

CJNE    A,#33H,FE4       ;IF NOT 3 CHECK 4,p
MOV     R0,#38H          ;MSB REF          3800H
MOV     R1,#00H          ;LSB REF
CALL    TARGET

CJNE    A,#34H,POSRET    ;IF NOT 4 CHECK FOR 'p'
MOV     R0,#4CH          ;MSB REF          4C00H
MOV     R1,#00H          ;LSB REF
CALL    TARGET

```

```

TXC1:MOV      DPTR,#TABL      ; TABLE START OF LOOKUP TBL
TXC:   MOV     A,@A+DPTR      ; BIN2HEX
      CLR     TI              ; CHS OUTPUT
      MOV     SBUF,a
      JNB    TI,$
      RET

      CJNE   A,#70H,RSH      ; IF NOT 'p' LOOK FOR NEXT CHS

      MOV    A,#'P'          ; P =
      CALL  TXC
      MOV    A,#' '
      CALL  TXC
      MOV    A,#'='
      CALL  TXC
      MOV    A,#' '
      CALL  TXC
      MOV    A,TH2
      ANL   A,#0F0H          ; MASK LOWER NIBBLE OF MSB
      SWAP  A
      CALL  TXC1
      MOV    A,TH2
      ANL   A,#0FH           ; MASK UPPER NIBBLE OF MSB
      CALL  TXC1

      MOV    A,TL2
      ANL   A,#0F0H          ; MASK LOWER NIBBLE OF LSB
      SWAP  A
      CALL  TXC1
      MOV    A,TL2
      ANL   A,#0FH           ; MASK UPPER NIBBLE OF LSB
      CALL  TXC1

      MOV    A,#'H'          ; H
      CALL  TXC
      MOV    A,#0AH          ; LF
      CALL  TXC
      MOV    A,#' '
      CALL  TXC
      MOV    A,#0DH          ; CR
      CALL  TXC

RSH:   LJMP  RDCHS           ; JMP TO RDCHS / BRIDGE REL JM

      CLR    TR0              ; disable t0
      CLR    TF0              ; clr flag
      MOV    TH0,#0F8h        ; >5mS delay
      MOV    TL0,#0h
      SETB  TR0               ; start t0
      JNB   TF0,$            ;
      RET

      SETB  P1.6              ; BOTH LED'S ON
      SETB  P1.7

      CLR   C
      MOV   A,R1              ; LSB POS COUNT
      SUBB A,TL2
      MOV   R3,A              ; LSB ERROR
      MOV   A,R0
      SUBB A,TH2
      MOV   R2,A              ; MSB ERROR
      JC   FMOD               ; ERROR -ve
      SETB P1.4              ; + DIR
      JMP  PROP

```

```

CLR C ;IF -ve MOD OF ERROR
MOV A,R3 ;LSB ERROR
SUBB A,#1H ;DEC R3
JNC COMPL ;LSB DEC SUFFICIENT
DEC R2 ;IF CARRY DEC MSB ALSO

CPL A
MOV R3,A ;MSB MOD ERROR

CPL MOV A,R2
A ;LSB MOD ERROR
MOV R2,A

MOV A,R2 ;MSB ERROR
JZ INV ;ERROR<255

SETB P1.5 ;RELEASE BRAKE
MOV A,CCAP0H
CJNE A,#0FFH,TARGET ;IF ALREADY RUNNING IGNORE

DEC A
MOV CCAP0H,A
CALL DELAY
CJNE A,#0H,DLOP ;INCREASE SPEED SLOWLY TILL 100PC
JMP TARGET ;AFTER START, TARGET

INV: SETB P1.5 ;REL BRAKE EVEN IF ERR<255
MOV A,#0FFH ;ERR<255, INV FOR PWM REG
CLR C
SUBB A,R3 ;INV OF PWM
MOV CCAP0H,A ;PWM='ERROR'
CJNE A,#0FAH,LEC

BRK: MOV CCAP0H,0FFH ;PWM=0
CLR P1.5 ;BRAKE ON
clr p1.7 ;led1 off
setb p1.6 ;led2 on

MOV A,#07H ;SOUND BELL
clr ti
mov sbuf,a
jnb ti,$

CALL DELAY

MOV A,#07H ;SOUND BELL
CLR TI
MOV SBUF,A
JNB TI,$

JMP RDCHS

LEC: JNC BRK ;IF ERROR < 5 BRAKE
JMP TARGET ;WAIT TILL ERROR<= 5

```

- INDEX ISR -----

```

SETB TR2 ;ENABLE T2
MOV TH2,#10H ;REF POS 1st INDEX =1000H
MOV TL2,#00H

MOV CCAP0H,0FFH ;PWM=0

SETB P1.6 ;BOTH LED'S ON
SETB P1.7

```

```

CALL      DELAY
SETB     P1.6                ;BOTH ON
SETB     P1.7

CALL      DELAY
CLR      P1.6                ;BOTH OFF
CLR      P1.7

CALL      DELAY
SETB     P1.6                ;BOTH ON
SETB     P1.7

CALL      DELAY
CLR      P1.6                ;BOTH OFF
CLR      P1.7

CALL      DELAY
SETB     P1.6                ;BOTH ON
SETB     P1.7

CLR      P1.5                ;BRAKE ON
clr      p1.7                ;led1 off
setb     p1.6                ;led2 on
MOV      IE,#10H            ;DISABLE FURTHER EXT INT, EN T2 INT
CLR      IE0                ;EXT 0 SOURCE
RETI

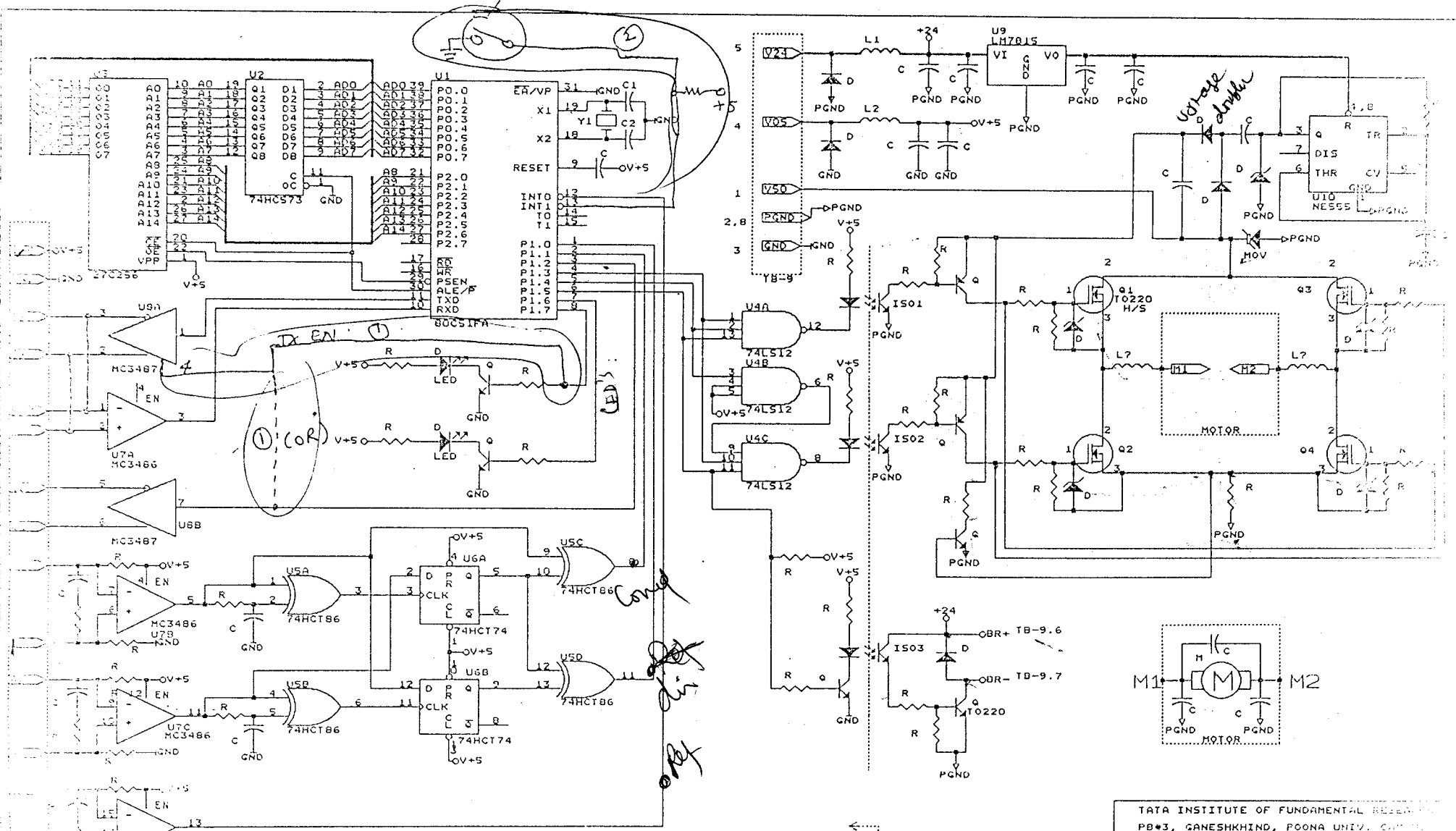
```

2 ISR -----

```

CLR      P1.5                ;BRAKE ON
clr      p1.7                ;led1 off
setb     p1.6                ;led2 on
mov      r4,#10d
mov      a,#45h            ;E
clr      ti
mov      sbuf,a
jnb     ti,$
clr      ti
dec     r4
cjne    r4,#0h,M1
clr     ri
CLR     TF2                ;CLR SOURCE FOR T2 INT
RETI

```



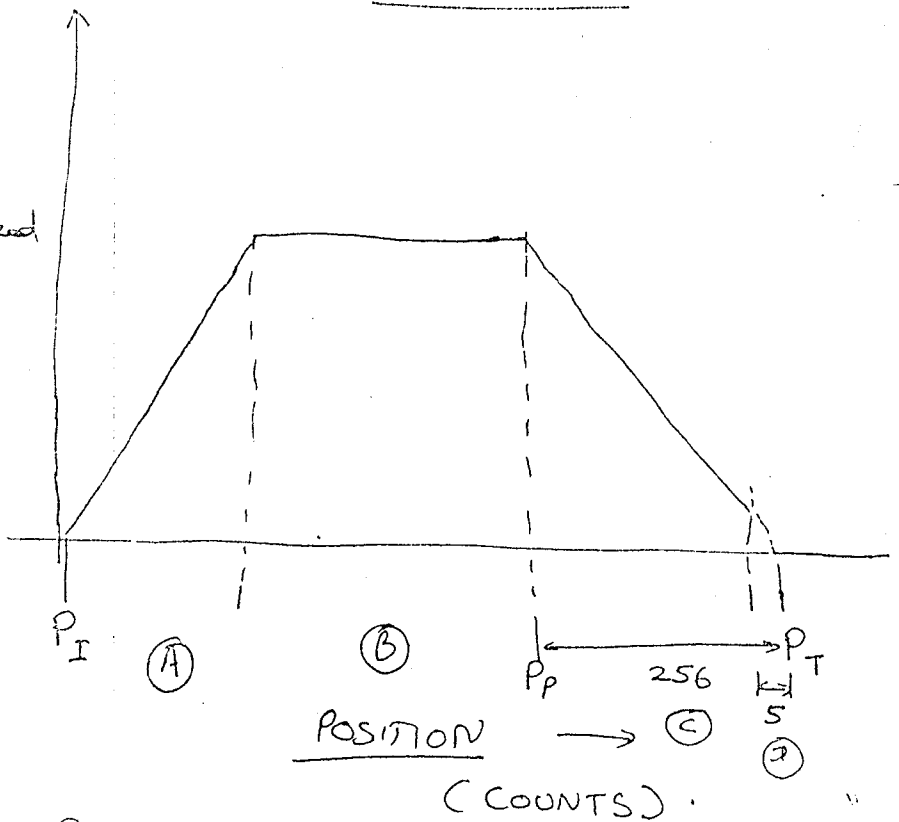
LOGIC GROUND--- GND
 POWER GROUND--- PGND

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MODIFICATIONS SUGGESTED

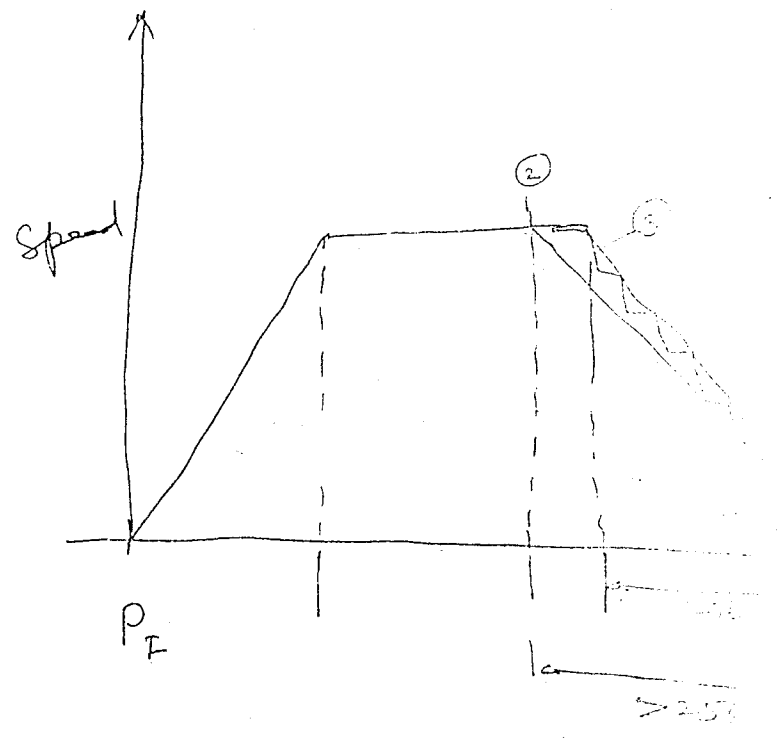
PRESENT



P_I - Initial Position
 P_P - Start of Proportional Region
 P_T - Target Position

Region A - Startup
 B - Position error > 256
 C - Prop. Region Pos error < 256 .

SOFTWARE



- ① Braking region can be changed to obtain proper final position.
 (Brake when Pos error > 256)
- ② Proportional region can be increased beyond 256 counts.
- ③ Proportional gain can be adjusted.